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Theoretical study of 2 × 2 planar array of equilateral triangular patch microstrip antenna on ferrite substrate

K K Verma* and K R Soni

Microwave Laboratory, Department of Physics,
Malaviya National Institute of Technology Jaipur-302 017, Rajasthan, India

E-mail kkvermaphd@rediffmail.com

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act This paper presents theoretical investigations of the radiation properties of a 2×2 element planar array of equilateral triangular patch antenna built up on a magnetized ferrite substrate $Ni_{1.062}Co_{0.02}Fe_{1.948}O_4$ in the presence of plasma medium. The far zone EM-mode and P-...ation fields are derived using vector wave function techniques and pattern multiplication approaches. The results are obtained both in plasma ...and in free space. Computation of maximum field intensity, half power beam width (HPBW), full null beam width (FNBW) and other antenna parameters such as radiation conductance, directivity and quality factor has been made for the plasma and the free space medium. ...ults of this theoretical analysis suggest that due to ferrite substrate this planar array antenna can be operated at lower frequency range and there ...ion of size of such antenna when designed on $NiCoFeO_4$ ferrite substrate

84 40 Ba, 52 40 Fd

Introduction

...cent past, microstrip antennas built up on ferrite substrate ...finding increasing attention to the numerous scientist and ...ers. Ferrite materials have a significant amount of ...ropy at microwave frequencies. This anisotropy gets ...ed by applying external d c magnetic field in ferrite mate-...ls and brings about non-reciprocal behavior in them.

More recently, with the availability of low loss commercial ...crowave, ferrite substrates have been used in microstrip ...enna applications to provide loading for antenna size ...uction. Microstrip antennas on pure dielectric substrate have ...n extensively analysed in these years due to their advan-...es over bulky antennas. The band width of such antennas in ... frequency range is typically of the order of 1 percent and ... have not tried at lower ultra high frequency (UHF) due to ...rge size considerations.

Such antennas when mounted on aerospace vehicles en-...nter plasma medium during their travel in space. It has been pointed out that the radiation properties of microstrip antennas in plasma medium are modified to a great extent due to the generation of electroacoustic waves in addition to electromagnetic waves [1-5].

This paper describes the radiation characteristics of a 2 × 2 planar array of equilateral triangular patch microstrip antenna built up on a ferrite substrate $Ni_{1.062}Co_{0.02}Fe_{1.948}O_4$. Design requirement and substrate characteristics considered for this analysis are listed in Table 1.

2. Radiation field expressions

The geometry and coordinate system of planar array antenna is shown in Figure 1. It consists of four identical triangular microstrip patch elements of arm length a, on a ferrite substrate $Ni_{1.062}Co_{0.02}Fe_{1.948}O_4$ of thickness h and substrate permittivity ε_r. The array elements which are positioned along x-axis are separated by a distance d_x and those along y-axis are separated by a distance d_y. Each patch can be excited by a microstrip transmission line connected to the edge or by a coaxial line from the back at the plane $\phi = 0$. We have considered the patch as

*Corresponding Author

Table 1. Design requirement and substrate characteristics of ferrite substrate $Ni_{1.062}Co_{0.02}Fe_{1.948}O_4$

Design frequency (f)	1.0 GHz
Relative permittivity (ε_r)	14.78
Dielectric loss tangent ($\tan\delta_\varepsilon$)	0.0005
Magnetic loss tangent ($\tan\delta_m$)	0.005
Applied d.c. magnetic bias field (H_0)	7.96 × 10^4 Amp/m
Saturation magnetization ($\mu_0 M_s$)	0.03 Tesla
Patch dimension (a)	0.022 meter
Substrate thickness (h)	0.0016 meter
Gyromagnetic ratio (γ)	1.76 × 10^{11} rad/sec Tesla
Separation between the array element along x- and y-axis respectively ($d_x = d_y$)	$\lambda/2$
Progressive phase excitation difference along x- and y-axis respectively (β_x, β_y)	$\pi/2$

a cavity which acts as a disc resonator. Among the various modes that may be excited in such disc resonator, we have considered TM_{mn} mode with respect to z-axis. Here, m and n are the mode numbers associated with x- and y-directions respectively.

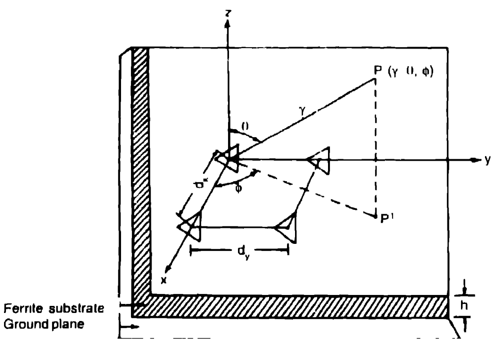


Figure 1. Configuration and coordinate system of 2×2 element planar array equilateral triangular patch array microstrip antenna on ferrite substrate

The E_z component of field inside the cavity for dominant mode is given as

$$E_z = A_{1,0}\left[2\cos\left(2\pi x/\sqrt{3}a + 2\pi/3\right)\cos(2\pi y/3a) + \cos(4\pi y/3a)\right] \quad (1)$$

A ferrite slab may support various guided modes and these propagating modes get affected in the presence of applied magnetic bias field. A strong bias field may be obtained easily along the plane of antenna by placing two pole pieces of magnet in close contact with substrate. When propagation of electromagnetic waves take place along this bias field in infinite ferrite having two orthogonal feeds, two planar modes namely left hand circularly polarized (LHCP) mode and right hand circularly polarized (RHCP) mode are generated.

The magnetic properties of ferrite substrate affect both modes, therefore the effective permeability for right hand circularly polarization and left hand circularly polarization can be given as [6-8]

$$\mu_{effR} = \mu_o\,\frac{1+\omega_m}{\omega_o - \omega},$$

$$\mu_{effL} = \mu_o\,\frac{1+\omega_m}{\omega_o + \omega},$$

where ω_0 and ω_m are the precession and forced precession frequencies, respectively and may be defined as

$$\omega_0 = \mu\gamma H_0,\ \omega_m = \mu\gamma H_s \text{ and } \omega = 2\pi f,$$

The expression of the resonant frequency ofconventional equilateral triangular microstrip antenna on ferrite substrate for TM_{10} mode is given as

$$f_r = \frac{2c\sqrt{\mu_0}}{3a\sqrt{\varepsilon_r\,\mu_{eff}}}$$

Other field components are obtained by solving Maxwell's equations. By image theory, the ground plane may be replaced by an image of the top conductor. The magnetic currents exist along the edges of triangular conductor and may be evaluated from

$$M = 2(E \times n),$$

where n is a unit vector normal to the aperture.

The total far zone fields of planar array antenna can be expressed by the fields of a single element positioned at origin multiplied by a factor which is referred to as the array factor. This method is widely known as pattern multiplication approach. Thus, using the expression of single element equilateral triangular patch microstrip antenna and neglecting the coupling between the elements, the total field of the array antenna can be expressed as [9-12]

E (total) = E (single element placed at the origin) × array factor (AF)

As the entire array is taken as uniform, the normalized form of the array factor (AF) is obtained and may be written as

$$AF_n(\theta,\phi) = \frac{1}{M}\times\frac{\sin(M\psi_x/2)}{\sin(\psi_x/2)}\times\frac{1}{N}\times\frac{\sin(N\psi_y/2)}{\sin(\psi_y/2)}$$

For the case of 2 × 2 element planar array of equilateral triangular patch microstrip antenna on ferrite substrate, the EM and P mode fields are given as

EM mode

$$E_{\theta t} = i\eta_0\omega\left[-F_x\sin\phi + F_y\cos\phi\right]\times\frac{1}{4}$$

$$\times\frac{\sin(\psi_x)}{\sin(\psi_x/2)}\times\frac{\sin(\psi_y)}{\sin(\psi_y/2)} \tag{8}$$

$$E_{\phi t} = -i\eta_0\omega\left[F_x\cos\theta\cos\phi + F_y\cos\theta\sin\phi\right]\times\frac{1}{4}$$

$$\times\frac{\sin(\psi_x)}{\sin(\psi_x/2)}\times\frac{\sin(\psi_y)}{\sin(\psi_y/2)} \tag{9}$$

$$E_{pt} = -2h\beta_p\omega_p^2/3a\omega\varepsilon_0\left(\omega^2-\omega_p^2\right)\times\exp\left(i\beta_p r\right)/r$$

$$\times\exp\left(i\beta_p a\sin\theta\cos\phi/\sqrt{3}\right)$$

$$\times\frac{1}{4}\times\frac{\sin\{\beta_p d_x\sin\theta\cos\phi+\beta_x\}}{\sin\{0.5(\beta_p d_x\sin\theta\cos\phi+\beta_x)\}}$$

$$\frac{\sin\{\beta_p d_y\sin\theta\cos\phi+\beta_y\}}{\sin\{0.5(\beta_p d_y\sin\theta\cos\phi+\beta_y)\}}\times|E_{px}+E_{py}|, \tag{10}$$

$\psi_x = \beta_e d_x\sin\theta\cos\phi+\beta_x$ and $\psi_y = \beta_e d_y\sin\theta\sin\phi+\beta_y$, M, N = elements placed along the x-axis and y-axis respectively, β_x, β_y = progressive phase excitation differences along x- and y-directions respectively, $E_{\theta t}$, $E_{\phi t}$ = components of total electric field vectors for EM-mode, Ept = total electric field vector for P-mode, F_x, F_y = vector electric potentials for x-component and y-component, respectively, E_{px}, E_{py} = P-mode electric field vectors for x-component and y-component respectively, β_e = phase propagation constant for EM- mode given by $2\pi A/\lambda_0$, β_p = phase propagation constant for P- mode given by $\beta_e c/v$, c = velocity of light, v = root mean square thermal velocity of electron, A = plasma frequency parameter given by $\left(1-\omega_p^2/\omega^2\right)^{1/2}$, ω_p = angular plasma frequency, ω = angular source frequency, η_0 = free space impedance equal to 120π ohms

3. Field patterns

The expression for total field pattern $R(\theta,\phi)$ is obtained as

$$R(\theta,\phi) = |E_{\theta t}|^2 + |E_{\phi t}|^2 \tag{11}$$

Then, the radiation field patterns in the E-plane ($\phi = 0$) and H-plane ($\phi = \pi/2$) are given as

$$R_e(\theta,\phi) = |E_{\theta t}|^2 + |E_{\phi t}|^2 = \eta_0^2\omega^2\left(|F_x|^2 + |F_x|^2\cos^2\theta\right),$$

(E-plane) (12)

$$R_h(\theta,\phi) = |E_{\theta t}|^2 + |E_{\phi t}|^2 = \eta_0^2\omega^2\left(|F_x|^2 + |F_x|^2\cos^2\theta\right) \tag{13}$$

The values of R_e and R_h are calculated for a case taking f= 1 GHz, a = 2.2 cm, ε_r = 14.78, Applied d.c. magnetic bias field (H_0) = 7.96 × 10^4 Amp/m, Saturation magnetization ($\mu_0 M_s$) = 0.03 Tesla, d_x = d_y = $\lambda/2$ and the phase difference $\beta_x = \beta_y = \pi/2$. The results are plotted in Figures 2 and 3 for two different planes ($\phi = 0$ and $\phi = \pi/2$) for A = 1.0 *i.e.* in free space and in Figures 4 and 5 for two different planes ($\phi = 0$ and $\phi = \pi/2$) A = 0.5 *i.e.* in plasma medium for TM_{10} mode of excitation. Pattern characteristics of planar array are computed in Table 2. Comparison between radiation properties of the planar array antenna on dielectric and ferrite substrate are shown in Table 3

Table 2. Pattern characteristics of the 2 × 2 element planar array of triangular microstrip patch antenna on ferrite substrate

Sr No	Pattern characteristics	$\phi = 0$ plane A = 1.0		$\phi = 0$ plane A = 0.5		$\phi = \pi/2$ plane A = 1.0		$\phi = \pi/2$ plane A = 0.5	
		RHCP	LHCP	RHCP	LHCP	RHCP	LHCP	RHCP	LHCP
1	Half-power beam width (HPBW)	60°	80°	140°	150°	60°	80°	140°	150°
2	Full null beam width (FNBW)	120°	140°	180°	280°	120°	140°	180°	280°
3	Direction of max radiation	90°	90°	180°	180°	90°	90°	180°	180°

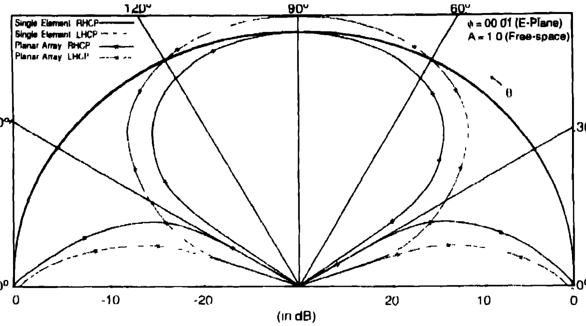


Figure 2. E plane radiation patterns of 2×2 element planar array and single-element equilateral triangular patch microstrip antennas on ferrite substrate for $A = 1\ 0$ (free space)

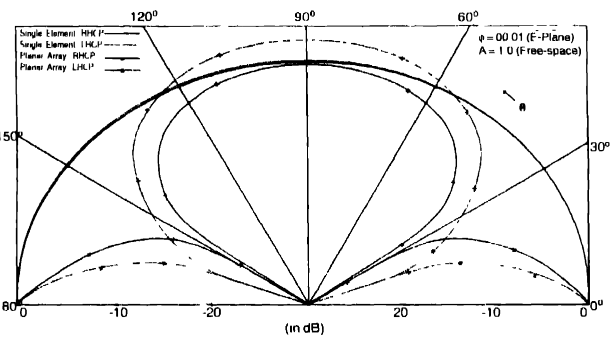


Figure 3. H-plane radiation patterns of 2×2 element planar array and single-element equilateral triangular patch microstrip antennas on ferrite substrate for $A = 1\ 0$ (free space)

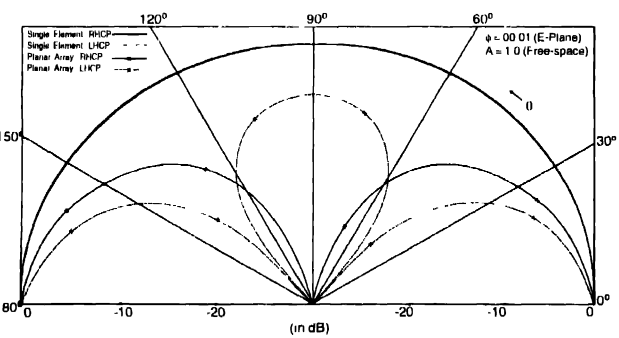


Figure 4. E-plane radiation patterns of 2×2 element planar array and single-element equilateral triangular patch microstrip antennas on ferrite substrate for $A = 0\ 5$ (plasma)

4. Other antenna parameters

4.1 Radiation conductance

The total power radiated can be calcu... by employing Poynting's theorem and ... as .

$$P_t = (½)\ (½)\ \left\{\mathrm{Re}\iint s\left(E \times H^*\right) \ldots\right.$$

The factor ½ is due to the fact that the p... is radiated through the upper half space and S is the total spherical surface area

$$P_t = \left(\tfrac{1}{4}\right)\left\{\mathrm{Re}\iint s\left(E_\theta H_\phi^* - E_\phi H_\theta \ldots\right.\right.$$

Thus, the expression for radiated p... EM-modes is obtained using the relation

$$P_t = (A/4\eta_0)\int_0^{2\pi}\int_0^{\pi}\left\{\left|E_{\theta t}\right.\right.$$

$$\times r^2 \sin\theta\ d\theta d\phi$$

Here, $A = \left(1 - \omega_p^2/\omega^2\right)^{1/2}$ and ω ... the plasma to source frequency ratio

The radiation conductance of an anten... EM-mode can be defined as

$$G_e = 2\ P_e/V_0^2$$

4 2 Directivity

The directivity of an antenna is defined ... ratio of the maxium radiation intensity ... per unit solid angle) to the average radi... intensity It can be expressed as

$$D_e = 4\ \pi\left\{\max\left(\left|E_{\theta t}\right|^2 + \left|E_{\phi t}\right|^2\right)\right\}$$

$$\left\{\int_0^{2\pi}\int_0^{\pi}\left|E_{\theta t}\right|^2 + \left|E_{\phi t}\right|^2 r^2 \sin\theta\ d\theta\ d\phi\right\}$$

4 3 Quality factor

A parameter specifying frequency selec... of a resonant circuit is the quality factor Q ... can be defined as the ratio between e... stored in the system and the energy lost

The total Q of a microstrip radiating el... comprises contributions due to the rad... Q_r, conductor loss Q_c, and dielectric lo... quality factors, so

Table 3. Comparison between radiation properties of the 2×2 element planar array of triangular microstrip patch antenna on pure dielectric substrate and ferrite substrate in TM_{10} mode

	Patch dimensions (meters)	Resonant frequency (GHz)	Radiation conduction (mho)	Quality factor	Directive gain (dB)
Pure-dielectric substrate ε_r = 2 32, h = 0 0016	0 022	5 97	0 003	7 756	6 518
Ferrite substrate, LHCP ε_r = 14 78 h = 0 0016	0 022	1 2	$1\ 116\times10^{-4}$	252 15	7 5
Ferrite substrate RHCP ε_r = 14 78 h = 0 0016	0 022	0 9	$7\ 356\times10^{-5}$	259 09	6 625

$$1/Q = 1/Q_r + 1/Q_c + 1/Q_d, \tag{17}$$

$$Q_r = \omega W_t / P_r,$$

$$Q_c = \omega W_t / P_c = (\pi f \mu \sigma)^{1/2} h,$$

$$Q_d = \omega W_t / P_d = 1/\tan\delta$$

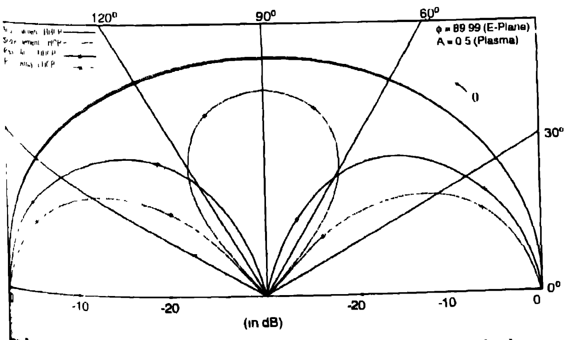
...ere W_t is the energy stored in the antenna element, P_c and ...wer loss factors due to the conductors and dielectric, ...ctively σ is the conductivity of the conductors

...nergy stored in the triangular radiating element is given

$$W_t = (\varepsilon h/2) \iint |E_z(xy)|^2 \, dx \, dy \tag{18}$$

...onclusion

...radiation characteristics of 2 × 2 planar array of equilateral ...gular patch microstrip antenna on ferrite substrate in plasma as well as in free space have been studied by considering presence of bias magnetic field in the direction of propagation of e m waves The results of the array geometry are compared with those of single-element of equilateral triangular patch microstrip antenna Figures 2 and 3 illustrate the radiation patterns of 2 × 2 element planar array of equilateral triangular microstrip antenna for two planes ($\phi = 0$ and $\phi = \pi/2$) in free space It is observed that the beam splits up into major and minor lobes and corresponding position of maximum radiation is shifted The radiation patterns of a single element antenna for both RHCP and LHCP waves contain only one major lobe of considerably wide beam width, while the array geometry for both RHCP and LHCP waves produces a directive beam with a narrow beam width The radiation field patterns of this array antenna indicate that beam width with LHCP waves is larger than that with RHCP waves. Figures 4 and 5 illustrate the radiation patterns of 2 × 2 element planar array of equilateral triangular microstrip antenna for two planes ($\phi = 0$ and $\phi = \pi/2$) in plasma medium (A=0 5) From these figures, it is observed that the shape of the field patterns has been changed and it redistributes the field intensity The computed values of pattern characteristics of the planar array are given in Table 2 It is clear from this table that half power beam width of the patterns in free space is relatively small in comparison to plasma medium Thus it reveals that the patterns are more directive in free-space than in plasma medium From Tables 4 and 5, it is clear that the effect of plasma on radiation conduction of single element as well as planar array of equilateral triangular microstrip antenna in RHCP mode is significantly weaker than the LHCP mode The directivity of 2×2 element planar array of equilateral triangular microstrip antenna with LHCP waves is little higher than



...re 5 H plane radiation patterns of 2x2 element planar array and single-elemen... ...teral triangular patch microstrip antennas on ferrite substrate for A = 0 5 (plasma)

Table 4. Antenna parameters for single element and 2 × 2 planar array equilateral triangular patch microstrip antenna o ferrite substrate with RHCP waves

Sr No	Plasma parameter (A)	G_t for single element (mho)	G_t for 2x2 planar array (mho)	D_t for single element (dB)	D_t for 2 x2 planar array	Q_t for single element	Q_t for 2x2 planar array
1	1.0	1.588×10^{-5}	7.356×10^{-5}	4.751	6.625	444.418	259.806
2	0.9	1.428×10^{-5}	6.819×10^{-5}	4.755	6.414	453.355	270.265
3	0.8	1.268×10^{-5}	6.152×10^{-5}	4.758	6.244	462.635	284.459
4	0.7	1.109×10^{-5}	5.379×10^{-5}	4.761	6.108	472.281	302.933
5	0.6	9.468×10^{-6}	4.537×10^{-5}	4.764	5.99	482.319	325.981
6	0.5	7.911×10^{-6}	3.673×10^{-5}	4.766	5.885	492.774	353.601
7	0.4	4.743×10^{-6}	2.829×10^{-5}	4.768	5.74	503.678	385.479
8	0.3	4.743×10^{-6}	2.073×10^{-5}	4.769	5.523	515.061	421.119
9	0.2	3.161×10^{-6}	1.309×10^{-5}	4.77	5.248	526.961	460.222
10	0.1	1.58×10^{-6}	6.379×10^{-6}	4.771	4.931	539.417	503.304
11	0.0	0.0	0.0	4.771	4.772	–	–

Table 5 Antenna parameters for single element and 2x2 planar array equilateral triangular patch microstrip antenna o ferrite substrate with LHCP waves

Sr No	Plasma parameter (A)	G_t for single element (mho)	G_t for 2x2 planar array (mho)	D_t for single element (dB)	D_t for 2 x2 planar array	Q_t for single element	Q_t for 2x planar array
1	1.0	2.811×10^{-5}	1.116×10^{-4}	4.736	7.5	449.784	252.15
2	0.9	2.526×10^{-5}	1.071×10^{-4}	4.742	7.151	462.144	258.296
3	0.8	2.242×10^{-5}	1.013×10^{-4}	4.748	6.767	475.146	266.627
4	0.7	1.959×10^{-5}	9.287×10^{-5}	4.754	6.475	488.847	279.728
5	0.6	1.678×10^{-5}	8.142×10^{-5}	4.758	6.24	503.311	299.766
6	0.5	1.397×10^{-5}	6.751×10^{-5}	4.762	6.063	518.612	328.353
7	0.4	1.117×10^{-5}	5.237×10^{-5}	4.765	5.921	534.831	366.356
8	0.3	8.37×10^{-6}	3.741×10^{-5}	4.768	5.737	552.062	413.675
9	0.2	5.578×10^{-6}	2.364×10^{-5}	4.77	5.435	570.413	469.492
10	0.1	2.788×10^{-6}	1.133×10^{-5}	4.771	5.024	590.006	533.852
11	0.0	0.0	0.0	4.771	4.77	–	–

that with RHCP waves for all the plasma frequency and free space. The total quality factor of planar array antenna with LHCP waves is lower than that with RHCP waves.

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